

Classical and Quantum Error Correction

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Prelude





Inside the box: Channel + Detector errors





Noisy memoryless channels



Simple memoryless channels

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• Binary erasure channel (BEC)

• Binary symmetric channel (BSC)

- Binary input additive white Gaussian noise (AWGN) channel, σ^2





 $1-\alpha$

 $1-\alpha$

α

0

α

0





Channel capacity – Binary Erasure Channel



We lose a fraction ε of the bits. How do we recover that data?



Let $x = [x_1, ..., x_n]$ be a codeword transmitted over a memoryless channel and let $y = [y_1, ..., y_n]$ be the corresponding channel output. Then the conditional probability density p(y|x) can be written as

- A. $p(\mathbf{y}|\mathbf{x}) = \sum_{i=1}^{n} p(y_i|x_i)$
- B. $p(\mathbf{y}|\mathbf{x}) = p(y_1|x_1) p(y_2|x_2) p(y_3|x_3) \cdots p(y_{n-1}|x_{n-1}) p(y_n|x_n)$
- C. $p(\mathbf{y}|\mathbf{x}) = p(y_1|x_2) + p(y_2|x_3) + p(y_3|x_4) + \dots + p(y_{n-1}|x_n)$
- *D.* $p(y|x) = p(y_1|x_2) \cup p(y_2|x_3) \cup p(y_3|x_4) \cup \dots \cup p(y_{n-1}|x_n)$
- *E.* $p(\mathbf{y}|\mathbf{x}) = p(\{y_1 + y_2 + , ... + y_n\} \cup \{x_1 + x_2 + \dots + x_n\})$
- F. None of the above











 $C = 1 - \varepsilon$: But the *n*-bit repetition code sends just 1 bit / *n* channel uses!







- Message: $m = [m_1, m_2, ..., m_k]$
- Codeword: $x = [x_1, x_2, ..., x_n]$
- Code rate: $R = \frac{k}{n} \le C$ (Capacity) ≤ 1
- Received word: $\mathbf{y} = [y_1, y_2, \dots, y_n]$
- The decoder tries to find \hat{x} (or \hat{m}) from y so that the probability of bit/codeword error is minimal.
- In other words, decoder tries to find a codeword that is "closest" to y.

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[n, k] binary linear codes

Generator matrix
$$(k \times n)$$
: $G = \begin{bmatrix} g_1 \\ g_2 \\ \vdots \\ g_k \end{bmatrix} \in \{0,1\}^{k \times n}$ (rank k binary matrix)

Encoding: $x = mG = m_1g_1 + m_2g_2 + \dots + m_kg_k \in \{0,1\}^n$ (XOR)

n-bit Repetition Code:
$$G = [g] = [1 \ 1 \ 1 \ \cdots \ 1]$$

 $m = [m] \xrightarrow{Encode} x = mG = [m \ m \ m \ \cdots \ m]$

[n = 5, k = 2] Code: (contains $2^k = 4$ codewords to encode $2^k = 4$ messages)

$$G = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 1 & 0 \end{bmatrix} \xrightarrow{m} = \begin{bmatrix} 0 & 0 \end{bmatrix} \xrightarrow{Encode}_{Encode} \begin{array}{c} x = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ \hline Encode \\ \hline m = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{array}{c} m = \begin{bmatrix} 0 & 1 \end{bmatrix} \xrightarrow{Encode}_{Encode} \begin{array}{c} x = \begin{bmatrix} 0 & 1 & 1 & 1 & 0 \end{bmatrix} \\ \hline x = \begin{bmatrix} 1 & 0 & 1 & 1 & 1 & 0 \end{bmatrix} \begin{array}{c} m = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{array}{c} \hline Encode \\ \hline Encode \\ \hline m = \begin{bmatrix} 1 & 1 \end{bmatrix} \end{array} \begin{array}{c} x = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \end{bmatrix} \\ \hline x = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \end{bmatrix} \end{array}$$



Parity-check matrix



[n = 5, k = 2] Code: (contains $2^k = 4$ codewords to encode $2^k = 4$ messages) $G = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 1 & 0 \end{bmatrix} = \begin{bmatrix} g_1 \\ g_2 \\ \vdots \\ g_k \end{bmatrix} \begin{array}{c} m = \begin{bmatrix} 0 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 0 & 1 \end{bmatrix} \\ \frac{Encode}{Encode} \\ m = \begin{bmatrix} 1 & 0 \end{bmatrix} \\ x = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{array}{c} m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 & 1 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 & 1 \end{bmatrix} \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 \\ m = \begin{bmatrix} 1 & 0 & 0 & 0 \\ m = \begin{bmatrix} 1 & 0 & 0 & 0 \\ m = \begin{bmatrix} 1 & 0 & 0 & 0 \\ m = \begin{bmatrix} 1 & 0 & 0 & 0 \\ m = \begin{bmatrix} 1 & 0 & 0 & 0 \\ m = \begin{bmatrix} 1 & 0$ Encoding: $x = [m_1 \ m_2]G = [m_1 \ m_2 \ m_1 + m_2 \ m_2 \ m_1]$ $= \begin{bmatrix} x_1 & x_2 & x_3 & x_4 & x_5 \end{bmatrix}$ Parity-checks: $s_1 = x_1 + x_2 + x_3 = m_1 + m_2 + (m_1 + m_2) = 0$ $s_2 = x_2 + x_4 = m_2 + m_2 = 0$ $(HG^T)m^T = 0 \Rightarrow HG^T = 0$ $s_3 = x_1 + x_5 = m_1 + m_1 = 0$ Syndrome

 Parity-check matrix
 1
 1
 0
 0

 $(n-k) \times n$:
 1
 0
 0
 1
 0
 0
 1
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For an [n, k] linear block code, what are the dimensions of the generator and paritycheck matrices?

- *A. G*: $kn \times n$ and $H: (n k)n \times n$
- *B. G*: $k \times n$ and $H: (n k) \times n$
- C. G: $k \times k$ and $H: (n-k) \times (n-k)$
- *D. G*: $n \times k$ and $H: n \times (n k)$







Correcting errors

 \mathbf{n}

 $\mathbf{\Omega}$

Δ

 $\mathbf{\Omega}$

Codebook

Channel

ΓΛ

$$y = \begin{bmatrix} 1 & E & E & 1 & 1 \end{bmatrix} \xrightarrow{Decode} \widehat{x} = \begin{bmatrix} 1 & 1 & 0 & 1 & 1 \end{bmatrix}$$

Channel = BSC: $e = \begin{bmatrix} 0 & 1 & 0 & 0 \end{bmatrix}$; $s = Hy^T = H(x + e)^T = He^T = \begin{bmatrix} 1 & 1 & 0 \end{bmatrix}^T$ $y = \begin{bmatrix} 1 & 0 & 0 & 1 & 1 \end{bmatrix} \xrightarrow{Decode} \hat{x} = y + e = \begin{bmatrix} 1 & 1 & 0 & 1 & 1 \end{bmatrix}$

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Protecting information by coding

















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Consider the code with the following parity-check matrix

$$H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \\ 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$

and assume that the channel output is $y = [1 \ 1 \ 0 \ 0 \ 0]$. The syndrome is then:

- $A. \quad s = \begin{bmatrix} 1\\0\\0 \end{bmatrix}$ $B. \quad s = \begin{bmatrix} 2\\1\\1 \end{bmatrix}$ $C. \quad s = 3$ $D. \quad s = 6$ $E. \quad s = 2^{3}$
- F. None of the above
- G. I'm not sure



Consider the code with the following parity-check matrix

$$H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \\ 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$

and assume that the channel output is $y = [0 \ 0 \ 0 \ 0 \ E]$, where *E* is the erasure symbol. The erasure decoder will produce the following vector as the output:

A.
$$x = [0 \ 0 \ 0 \ 0 \ 0]$$

B. $x = [0 \ 0 \ 0 \ 0 \ 1]$
C. $e = [0 \ 0 \ 0 \ 0 \ 1]$
D. $m = [0 \ 0 \ 1]$
E. $m = [0 \ 1]$

- F. None of the above
- G. I'm not sure







Minimum distance





Hamming distance = the number of positions in which two binary vectors differ Minimum distance = the Hamming distance between two closest codewords









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Hamming distance = the number of positions in which two binary vectors differ Minimum distance d = the Hamming distance between two closest codewords

- = the Hamming distance between x_1 and x_2
- = the Hamming distance between x_1 and x_3
- = the Hamming distance between x_2 and x_4
- = the Hamming distance between x_3 and x_4

= 3

The code encodes k = 2 message bits into n = 5 code bits with distance d = 3

Hamming spheres of radius $t = \frac{d-1}{2} = 1$ around codewords don't intersect Code corrects t = 1 error or d - 1 = 2 erasures; detects d - 1 = 2 errors



Minimum distance



$\binom{n}{t} = 5$ vectors at Hamming distance t = 1 from any codeword



Hamming distance = the number of positions in which two binary vectors differ Minimum distance = the Hamming distance between two closest codewords







Dual code C^{\perp}



Generator
and
Parity-check $G = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 1 & 0 \end{bmatrix}$ $H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \\ 1 & 0 & 0 & 0 & 1 \end{bmatrix}$ Syndrome
Syndrome

 1

Duality (Orthogonality): $GH^T = 0$

Dual Code Row space of *H*: $x_i v_i^T = 0$

Code C

Coc

Coc



 Λ Λ



Protecting information by coding

























A linear block code with minimum distance *d* can correct any

- A. weight (d 1) errors B. weight - $\left(\frac{d-1}{2}\right)$ errors C. weight - $\left(\frac{d}{2}\right)$ errors D. weight - $\left(\frac{d}{2} + 1\right)$ errors
- E. I'm not sure





[7,4,3] Hamming code





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Num Networks NSF-ERC Example – correcting single errors

- Rewrite *H* using column vectors $H = [h_1, h_2, \dots, h_j, \dots, h_n]$
- Error vector $e = [e_1, e_2, ..., e_j, ..., e_n]$
- Syndrome $s^{T} = He^{T} = [e_{1}h_{1}, e_{2}h_{2}, \dots, e_{j}h_{j}, \dots, e_{n}h_{n}]$
- Suppose *e* contains only one binary 1 at the *j*-th position, i.e., $e = [0, 0, \dots, e_j = 1, \dots 0]$
- Then $s^{\mathrm{T}} = \mathsf{h}_j$

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- In order to correct a single error in the codeword, the columns of *H* must be all different and nonzero.
- The dimensions of *H* are $(n k) \times n$, thus the largest code length is $n = 2^{n-k} 1$.
- Thus, in this case, $k = n \log_2(n+1)$.







$$H = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 0 \\ 0 & 1 & 1 & 1 \end{bmatrix}$$
$$G = \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \\ 1 & 1 & 1 \\ 1 & 0 & 1 \end{bmatrix} \begin{pmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

 \mathbf{O}

 $\left(\right)$

 $g_i H^{\mathsf{T}} = 0$ for any row g_i of G $GH^{\mathsf{T}} = 0$









Perfect Code: The $2^k = 16$ Hamming spheres cover all the $2^n = 128$ vectors!



The channel introduces random bit flips but in any 10 consecutive bits, it introduces no more than a single bit flip. You choose to use a Hamming code to protect user information bits against bit flips in such channel. The largest number of information bits that can be protected by encoding them using the Hamming code is:

- A. 0
- B. 1
- C. 2
- D. 3
- E. 4
- F. 5
- G. 9

H. I'm not sure




- d = d(x, y): the Hamming distance between x and y
- For the BSC channel: $P(y|x) = \alpha^d (1-\alpha)^{n-d}$
- ML decoding rule:

$$\hat{x} = \underset{x \in C}{\operatorname{argmax}} P(x|y) = \underset{x \in C}{\operatorname{argmax}} \frac{P(x)P(y|x)}{P(y)}$$

- If all codewords are equally likely, then maximize $\log_2 P(y|x)$





• ML decoding rule:

 $\hat{x} = \underset{x \in C}{\operatorname{argmax}} \log_2 P(y|x)$

• For the BSC:

$$P(y|x) = \alpha^d (1 - \alpha)^{n-d}$$

$$\log_2 P(y|x) = d \log_2 \alpha + (n - d) \log_2 (1 - \alpha)$$

$$= d \log_2 \frac{\alpha}{1 - \alpha} + n \log_2 (1 - \alpha)$$

negative for $\alpha < 1/2$ independent of *d*

• Hence, \widehat{x} is the codeword closest to y:

$$\hat{x} = \operatorname*{argmin}_{x \in \mathcal{C}} d(x, y)$$







Example [6,3,3] code











Standard array decoding

	ML d	ecoding	g: $\hat{x} =$	$\operatorname*{argmin}_{x \in \mathcal{C}}$	d(x,y)	H =	$\begin{bmatrix} 1 & 1 \\ & 1 & 1 \\ 1 & & 1 \end{bmatrix}$	1 1 1	
				2^k					
	000 000	001 011	010 110	100 101	011 101	101 110	110 011	111 000	000
	000 001	001 010	010 111	100 100	011 100	101 111	110 010	111 001	001
-	000 010	001 001	010 100	100 111	011 111	101 100	110 001	111 010	010
2^{n-k}	000 100	001 111	010 010	100 001	011 001	101 010	110 111	111 100	100
	001 000	000 011	011 110	101 101	010 101	100 110	111 011	110 000	011
	010 000	011 011	000 110	110 101	001 101	111 110	100 011	101 000	110
	100 000	101 011	110 110	000 101	111 101	001 110	010 011	011 000	101







 2^k



ML decoding:	$\hat{x} = \operatorname{argmin} d(x, y = 110101)$
	$x{\in}\mathcal{C}$

000 000	001 011	010 110	100 101	011 101	101 110	110 011	111 000	00
000 001	001 010	010 111	100 100	011 100	101 111	110 010	111 001	00
000 010	001 001	010 100	100 111	011 111	101 100	110 001	111 010	01
000 100	001 111	010 010	100 001	011 001	101 010	110 111	111 100	10
001 000	000 011	011 110	101 101	010 101	100 110	111 011	110 000	01
010 000	011 011	000 110	<u>110 101</u>	001 101	111 110	100 011	101 000	11
100 000	101 011	110 110	000 101	111 101	001 110	010 011	011 000	1(
	000 000 000 001 000 010 000 100 001 000 010 000 100 000	000 000001 011000 001001 010000 010001 001000 100001 111001 000000 011010 000011 011100 000101 011	000 000001 011010 110000 001001 010010 111000 010001 001010 100000 100001 111010 010001 000000 011011 110010 000011 011000 110100 000101 011110 110	000 000001 011010 110100 101000 001001 010010 111100 100000 010001 001010 100100 111000 100001 111010 010100 001001 000000 011011 110101 101010 000011 011000 110110 101100 000101 011110 110000 101	000 000001 011010 110100 101011 101000 001001 010010 111100 100011 100000 010001 001010 100100 111011 111000 100001 111010 010100 001011 001001 000000 011011 110101 101010 101010 000011 011000 110110 101001 101100 000101 011110 110000 101111 101	000 000001 011010 110100 101011 101101 110000 001001 010010 111100 100011 100101 111000 010001 001010 100100 111011 111101 100000 100001 111010 010100 001011 001101 010001 000000 011011 110101 101010 110100 110010 000011 011000 110110 101001 101111 110100 000101 011110 110000 101111 110	000 000001 011010 110100 101011 101101 110110 011000 001001 010010 111100 100011 100101 111110 010000 010001 001010 100100 111011 111101 100110 001000 100001 111010 010100 001011 001101 010110 001000 100001 111010 010100 001011 001101 010110 111001 000000 011011 110101 101010 101100 011010 000011 011000 110110 101001 101110 0011100 000101 011110 110000 101111 101001 110100 000101 011110 110000 101111 101010 011	000 000 001 011 010 110 100 101 011 101 101 110 110 011 111 000 000 001 001 010 010 111 100 100 011 100 101 111 110 010 111 001 000 010 001 001 010 100 100 111 011 111 101 100 111 001 000 010 001 001 010 100 100 111 011 111 101 001 111 010 000 100 001 111 010 001 100 011 011 010 110 001 111 010 000 100 001 111 010 001 011 001 101 010 110 111 111 000 001 000 000 011 011 110 101 101 100 110 111 011 110 000 010 000 011 011 000 110 110 101 001 101 101 000 101 001 101 000 100 000 101 011 110 100 000 101 111 101 001 011 011 000 100 000 101 011 110 100 000 101 111 010 011 000

Complexity scales exponentially!!

Poll Question 7

The standard array of a linear block code is given below.

00000	10110	01101	11011
10000	00110	11101	01011
01000	11110	00101	10011
00100	10010	01001	111111
00010	10100	01111	11001
00001	10111	01100	11010
00011	10101	01110	11000
10001	00111	11100	01010

The word received from the channel is $y = [0\ 1\ 1\ 0\ 0]$, and the standard array decoder is used to estimate the transmitted codeword x. The estimated codeword is:

A. The standard array decoder cannot correct this error pattern

B.
$$x = [0 \ 0 \ 0 \ 0 \ 1]$$

- *C.* $x = [0 \ 0 \ 1 \ 0 \ 1]$
- *D.* $x = [0\ 1\ 1\ 0\ 0]$
- *E.* $x = [0 \ 1 \ 1 \ 0 \ 1]$
- *F.* $x = [0 \ 1 \ 1 \ 0 \ 0]$
- G. I'm not sure





Codes on Graphs and Iterative Decoding









Low-density parity-check (LDPC) codes 🕏

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- Linear block codes defined by sparse bipartite graphs
- The *Tanner* graph of an LDPC code *C* is a bipartite graph *G* with two sets of nodes:
 - the set of variable nodes
 - and the set of check nodes

$$V = \{1, 2, \dots, n\} \\ C = \{1, 2, \dots, m\}$$











- The check nodes (resp. variable nodes) connected to a variable node (resp. check node) are its "*neighbors*".
- The set of neighbors of a node $u\,$ is denoted by $\mathcal{N}(u)$
- The degree d_u of a node u is the size of $\mathcal{N}(u)$









- A vector $\mathbf{v} = [v_1, v_2, \dots, v_n]$ is a codeword if and only if for each check node, the modulo two sum of its neighbors (i.e., respective bits of the vector) is zero
- An (n, γ, ρ) regular LDPC code has a Tanner graph with n variable nodes each of degree γ and $m = n\gamma/\rho$ check nodes each of degree ρ
- This code has length n and rate $r = \frac{k}{n} \ge 1 \frac{\gamma}{n}$
- The Tanner graph is not uniquely defined by the code
- Each parity-check matrix produces one Tanner graph

A regular ($n = 25, \gamma = 3, \rho = 5$) LDPC code





Poll Question 8

The parity check matrix that corresponds to the following Tanner graph is



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$$A. H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 1 & 1 & 1 & 1 \\ 1 & 1 & 0 & 1 & 1 \\ 1 & 1 & 0 & 1 & 1 \end{bmatrix}$$
$$B. H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 & 1 \\ 1 & 1 & 0 & 1 & 1 \\ 1 & 1 & 0 & 1 & 1 \end{bmatrix}$$
$$C. H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 \\ 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$
$$D. H = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 1 \\ 1 & 1 & 0 & 0 & 1 \end{bmatrix}$$

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Poll Question 9

For the code given by the Tanner graph below, the following statement is false:



- *A.* n = 25 and $\rho = 5$
- B. The check degree is three
- *C.* n = 25 and the Tanner graph is bipartite
- D. It is a regular LDPC code





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Iterative decoders for BEC





Iterative decoding on BEC

















- a check involving a <u>single</u> erased bit
- other check















































Success !







Another example BEC simulation - 1









Another example BEC simulation - 2









BEC simulation - final



Stuck !





Decoding failures



 A BEC iterative decoder fails to converge to a codeword (correct or wrong) if at any iteration there is no check node connected to at most one erased variable node.



 Graph induced by a subset of check nodes each connected to at least two erased variables is a <u>stopping set</u>.





Gallager A/B algorithm



- The Gallager A/B algorithms are hard decision decoding algorithms in which all the messages are binary
- $|\varpi_{*\to i} = m|$ number of incoming messages to i which are equal to $m \in \{0, 1\}$. Associated with every decoding round k and variable degree d_i is a threshold b_{k,d_i} .
- The Gallager B algorithm is defined as follows:

$$\begin{split} \omega_{i \to \alpha}^{(0)} &= y_i \\ \varpi_{\alpha \to i}^{(k)} &= \left(\sum_{j \in \mathcal{N}(\alpha) \setminus i} \omega_{j \to \alpha}^{(k-1)} \right) \mod 2 \\ \omega_{i \to \alpha}^{(k)} &= \left\{ \begin{array}{ll} 1, & \text{if } | \varpi_{* \setminus \alpha \to i}^{(k)} = 1 | \ge b_{k, d_i} \\ 0, & \text{if } | \varpi_{* \setminus \alpha \to i}^{(k)} = 0 | \ge b_{k, d_i} \\ y_i, & \text{otherwise} \end{array} \right. \end{split}$$













iteration 1 – initialization all variables send zero

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iteration 1 – the second half















recall what messages were sent to variable nodes









iteration 2 – first half variables send the majority of incoming messages

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iteration 2 – second half




Iterations of Gallager B





iteration 2 - decision syndrome mismatch

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Iterations of Gallager B



messages sent to variable nodes in previous iteration





Iterations of Gallager B





iteration 3 – first half as when we started



()













Poll Question 10

The Gallager-B decoder on the BSC channel with cross over probability α operates by sending messages between variable and check node processing units. After receiving all three messages from its neighboring checks, assuming that the channel value is 0, the variable node processing unit of the variable shown in the picture below will send the following message to the check node processing unit of the remaining check:









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Decoding by belief propagation





Crossword puzzles



• Iterate!



Across:

4 Animal with long ears and a short tail.10 Person who is in charge of a country.12 In no place.

Down:

- 5 Pointer, weapon fired from a bow.
- 6 Accept as true.
- 7 A place to shoot at; objective.







• ML decoding rule:

$$\hat{x} = \operatorname*{argmax}_{x \in C} P(x|y)$$

- Must evaluate posterior for each of the 2^k codewords!
- Make bit-wise decisions instead:

$$\hat{x}_j = \underset{x_j \in \{0,1\}}{\operatorname{argmax}} P(x_j|y) = \underset{x_j \in \{0,1\}}{\operatorname{argmax}} \sum_{x_1,\dots,x_{j-1},x_{j+1},\dots,x_n \in \{0,1\}^{n-1}} P(x|y)$$

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Bit-wise maximum likelihood (ML) decoding



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Sum-product algorithm (SPA)



Belief propagation (BP)



Variable node (VN) update:

$$\mu_{x \to f}(x) = \prod_{h \in n(x) \setminus \{f\}} \mu_{h \to x}(x)$$

Check node (CN) update:

$$\mu_{f \to x}(x) = \sum_{\sim \{x\}} \left(f(X) \prod_{h \in n(f) \setminus \{x\}} \mu_{y \to f}(y) \right)$$

Variable node (VN) decision:

$$g_i(x_i) = \prod_{h \in n(x_i)} \mu_{h \to x_i}(x_i)$$



Decoders for channels with soft outputs

 In addition to the channel value, a measure of bit reliability is also provided



• Bit log-likelihood ratio (LLR) given y_i:

$$\lambda(x_i) = \log \frac{P(x_i = 0|y_i)}{P(x_i = 1|y_i)}$$

$$= \log \frac{\frac{p(y_i|x_i=0)P(x_i=0)}{p(y_i)}}{\frac{p(y_i|x_i=1)P(x_i=1)}{p(y_i)}} = \log \frac{p(y_i|x_i=0)P(x_i=0)}{p(y_i|x_i=1)P(x_i=1)}$$

$$= \log \frac{p(y_i | x_i = 0)}{p(y_i | x_i = 1)} + \log \frac{P(x_i = 0)}{P(x_i = 1)}$$

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- Log-likelihood ratio (LLR)
- Without prior knowledge on x_i :

$$\gamma_i = \lambda(x_i) = \log \frac{p(y_i | x_i = 0)}{p(y_i | x_i = 1)}$$

• For AWGN $(y_i = x_i + n_i; n_i \sim N(0,1))$:

$$\gamma_i = \log \frac{p(y_i | x_i = 0)}{p(y_i | x_i = 1)} = \frac{1}{2\sigma^2} (-(y_i - 1)^2 + (y_i + 1)^2) = \frac{y_i}{2\sigma^2}$$

• For BSC with parameter α :

$$\gamma_i = \begin{cases} \log \frac{1-\alpha}{\alpha} & \text{if } y_i = 0\\ \log \frac{1-\alpha}{\alpha} & \text{if } y_i = 1 \end{cases}$$



ullet





- The update rule
 $$\begin{split}
 \omega_{i \to \alpha}^{(0)} &= \gamma_i \\
 \varpi_{\alpha \to i}^{(k)} &= 2 \tanh^{-1} \left(\prod_{j \in \mathcal{N}(\alpha) \setminus i} \tanh\left(\frac{1}{2}\omega_{j \to \alpha}^{(k-1)}\right) \right) \\
 \omega_{i \to \alpha}^{(k)} &= \gamma_i + \sum_{\delta \in \mathcal{N}(i) \setminus \alpha} \varpi_{\delta \to i}^{(k)} \end{split}$$
- The result of decoding after k iterations, denoted by $\mathbf{x}^{(k)}$ is determined by the sign of

$$m_i^{(k)} = \gamma_i + \sum_{\alpha \in \mathcal{N}(i)} \varpi_{\alpha \to i}^{(k)}$$

If $m_i^{(k)} > 0$ then $x_i^{(k)} = 0$, otherwise $x_i^{(k)} = 1$



The min-sum approximation (MSA)





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Poll Question 11

The min-sum decoder operates by sending messages between variable and check node processing units. After receiving all three messages from its neighboring checks, assuming that the channel value is -3, the variable node processing unit of the variable shown in the picture below will send the following message to the check node processing unit of the remaining check:

A. sign(+3)sign(-2)sign(-2)min(|+3|, |-2|, |-2| = +2

- B. sign(+3)sign(-2)sign(+2)min(|+3|, |-2|, |+2| = -2)
- C. -1 + sign(+3)sign(-2)sign(-2)min(|+3|, |-2|, |-2| = -3 + 2 = -1
- *D.* -1 + sign(+3)sign(-2)sign(+2)min(|+3|, |-2|, |+2| = -3 2 = -5
- E. ∞
- F. -∞
- *G.* 0
- H. -4
- I. +4
- J. I'm not sure





Applications of LDPC codes



- Wireless networks, satellite communications, deep-space communications, power line communications
- Magnetic hard disk drives, optical communications, flash memories
- Standards include:
 - Digital video broadcast over satellite (DVB-S2 Standard) and over cable (DVB-C2 Standard), terrestrial television broadcasting (DVB-T2, DVB-T2-Lite Standards)
 - GEO-Mobile Radio (GMR) satellite telephony (GMR-1 Standard), local and metropolitan area networks (LAN/MAN) (IEEE 802.11 (WiFi))
 - Wireless personal area networks (WPAN) (IEEE 802.15.3c (60 GHz PHY)), wireless local and metropolitan area networks (WLAN/WMAN) (IEEE 802.16 (Mobile WiMAX)
 - Near-earth and deep space communications (CCSDS), wire and power line communications (ITU-T G.hn (G.9960))
 - Ultra-wide band technologies (WiMedia 1.5 UWB)







Quantum Fundamentals





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A qubit is a 2-dimensional vector Networks

Computational basis states: "Ket 0", "Ket 1"



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 $|1\rangle$



 $|0\rangle$

 $|1\rangle$

Rotating to the conjugate basis

 $|\psi\rangle$

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- Conjugate basis states: "Ket +", "Ket -"
 - **Dirac notation:** $|+\rangle = \frac{1}{\sqrt{2}} \begin{bmatrix} 1\\1 \end{bmatrix}, \ |-\rangle = \frac{1}{\sqrt{2}} \begin{bmatrix} 1\\-1 \end{bmatrix}$
- A single-qubit state:

 $\begin{aligned} |\psi\rangle &= \alpha \left| 0 \right\rangle + \beta \left| 1 \right\rangle \\ &= \gamma \left| + \right\rangle + \delta \left| - \right\rangle \end{aligned}$

Bloch Sphere Visualizing 1 qubit

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What can you do with a qubit?

- Unitary operations: complex rotations, reversible
 - $U \in \mathbb{U}^{2 \times 2}$: $U^{-1} = U^{\dagger}$ Hermitian transpose

 $|1\rangle$



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Single Qubit: Unitary Operations



First quantum operation – Hadamard "gate"

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Z

 $|1\rangle$

• Switches between computational and conjugate bases

$$H |0\rangle = |+\rangle , H |1\rangle = |-\rangle$$
$$H |+\rangle = |0\rangle , H |-\rangle = |1\rangle$$

• Matrix representation:

$$H = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1\\ 1 & -1 \end{bmatrix}$$

Take any initial state

П

First quantum operation – Hadamard "gate"

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• Switches between computational and conjugate bases

$$H |0\rangle = |+\rangle , H |1\rangle = |-\rangle$$
$$H |+\rangle = |0\rangle , H |-\rangle = |1\rangle$$

• Matrix representation:

$$H = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1\\ 1 & -1 \end{bmatrix}$$
$$H^{-1} = H^{\dagger} = H$$

Take any initial state Rotate 90° by *Y* axis



First quantum operation – Hadamard "gate"

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• Switches between computational and conjugate bases

$$H |0\rangle = |+\rangle , H |1\rangle = |-\rangle$$
$$H |+\rangle = |0\rangle , H |-\rangle = |1\rangle$$

• Matrix representation:

$$H = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1\\ 1 & -1 \end{bmatrix}$$
$$H^{-1} = H^{\dagger} = H$$

Take any initial state Rotate 90° by Y axis Then rotate 180° by X axis



 $|1\rangle$

Poll Question 12

If we start with the single qubit state $|0\rangle$ and apply the *H* gate twice, what is the resulting state?

- A. $|+\rangle$
- B. $|-\rangle$
- *C.* |0>
- D. |1>
- E. None of the above
- F. I'm not sure









• The single-qubit Pauli matrices are: $(i = \sqrt{-1})$

$$I = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, X = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, Z = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, Y = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix}$$

Ζ

 $|1\rangle$

X

• These are π -rotations: $R_Z(\theta) = e^{-\frac{i\theta}{2}Z}$

$$e^{-\frac{i\pi}{2}Z} = \begin{bmatrix} e^{-\frac{i\pi}{2}} & 0\\ 0 & e^{\frac{i\pi}{2}} \end{bmatrix}$$
$$= e^{-\frac{i\pi}{2}} \begin{bmatrix} 1 & 0\\ 0 & e^{i\pi} \end{bmatrix}$$
$$= Z$$

Global phases don't matter!









The single-qubit Pauli matrices are: $(i = \sqrt{-1})$ ullet

$$I = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, X = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, Z = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, Y = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix}$$

Z

 $|1\rangle$

X

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These are π -rotations: $R_Z(\theta) = e^{-\frac{i\theta}{2}Z}$ ullet

$$e^{-\frac{i\pi}{2}Z} = \begin{bmatrix} e^{-\frac{i\pi}{2}} & 0\\ 0 & e^{\frac{i\pi}{2}} \end{bmatrix}$$
$$= e^{-\frac{i\pi}{2}} \begin{bmatrix} 1 & 0\\ 0 & e^{i\pi} \end{bmatrix}$$
$$\equiv Z$$
$$Z |+\rangle = |-\rangle, \ Z |-\rangle = |-\rangle$$









• The single-qubit Pauli matrices are: $(i = \sqrt{-1})$

$$I = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, X = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, Z = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, Y = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix}$$

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• Bit- and Phase-flip operations:

$$X \left| 0 \right\rangle = X \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \left| 1 \right\rangle$$

$$Z |0\rangle = |0\rangle , Z |1\rangle = -|1\rangle$$

 $Y = \imath XZ$ (Bit-Phase flip)



 $|1\rangle$

Ζ







• The single-qubit Pauli matrices are: $(i = \sqrt{-1})$

$$I = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, X = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, Z = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, Y = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix}$$

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• Bit- and Phase-flip operations:

$$Z \mid + \rangle \propto Z \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} 1 \\ -1 \end{bmatrix} = \mid - \rangle$$
$$X \mid + \rangle = \mid + \rangle, X \mid - \rangle = - \mid - \rangle$$

 $Y = \imath XZ$ (Bit-Phase flip)



 $|1\rangle$

Ζ







- How can we implement an arbitrary unitary operation?
- Classical Computing: NAND and NOR are universal
- Quantum Computing: A finite but universal gate set?





Universal gate set for one qubit

• The single-qubit Pauli gates are: $(i = \sqrt{-1})$

$$I = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, X = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, Z = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, Y = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix}$$

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• Hadamard gate:

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$$H = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1\\ 1 & -1 \end{bmatrix} = \frac{X+Z}{\sqrt{2}}$$

• *T* gate (π /4-rotation):

$$T = \begin{bmatrix} 1 & 0\\ 0 & e^{i\pi/4} \end{bmatrix} \equiv e^{-\frac{i\pi}{8}Z}$$

$$T^4 = Z, HZH = X, Y = \imath XZ$$



 $|1\rangle$

Ζ

Poll Question 13

The Phase gate is defined by $P = \sqrt{Z} = T^2$. What is the matrix representation of the Phase gate?

- $A. P = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}$ $B. P = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}$ $C. P = \begin{bmatrix} 1 & 0 \\ 0 & -i \end{bmatrix}$ $D. P = \begin{bmatrix} 1 & 0 \\ 0 & i \end{bmatrix}$
- E. None of the above
- F. I'm not sure





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Single Qubit: Measurements







Ζ

 $|1\rangle$

 ψ

Projective measurement

• Measure Z on $|\psi\rangle = \alpha |0\rangle + \beta |1\rangle$; "Bra psi" $\langle \psi | = |\psi\rangle^{\dagger}$ 1. First, diagonalize the measured "observable"

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 $Z = \left| 0 \right\rangle \left\langle 0 \right| - \left| 1 \right\rangle \left\langle 1 \right|$

2. Define projectors from eigenvectors

$$M_{+1} = |0\rangle \langle 0| , M_{-1} = |1\rangle \langle 1|$$

3. Possible outcomes "+1", "-1"

$$\mathbb{P}[+1] = \langle \psi | M_{+1} | \psi \rangle = |\alpha|^2$$
$$\mathbb{P}[-1] = \langle \psi | M_{-1} | \psi \rangle = |\beta|^2$$





Projective measurement

• Measure Z on $|\psi\rangle = \alpha |0\rangle + \beta |1\rangle$; "Bra psi" $\langle \psi | = |\psi\rangle^{\dagger}$ 1. First, diagonalize the measured "observable"

 $|0\rangle$

 $|1\rangle$

 $Z = \left| 0 \right\rangle \left\langle 0 \right| - \left| 1 \right\rangle \left\langle 1 \right|$

2. Define projectors from eigenvectors

$$M_{+1} = |0\rangle \langle 0| , M_{-1} = |1\rangle \langle 1|$$

3. Possible outcomes "+1", "-1"

$$\mathbb{P}[+1] = \langle \psi | M_{+1} | \psi \rangle = |\alpha|^2$$
$$\mathbb{P}[-1] = \langle \psi | M_{-1} | \psi \rangle = |\beta|^2$$

4. Post-measurement state

$$|\psi_{\pm}\rangle = \frac{M_{\pm 1} |\psi\rangle}{\sqrt{\mathbb{P}[\pm 1]}} = |0/1\rangle$$

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$$\begin{split} |\psi\rangle &= \alpha \, |0\rangle + \beta \, |1\rangle \\ \alpha, \beta \in \mathbb{C}, \ |\alpha|^2 + |\beta|^2 = 1 \end{split}$$

Technically, a qubit can store infinite information!

But NO measurement can retrieve it exactly!

This is true INDEPENDENT of the measurement basis







Projective measurement

 $|0\rangle$

 $|1\rangle$

 $|\psi|$

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- Measure *X* on $|\psi\rangle = \alpha |0\rangle + \beta |1\rangle = \gamma |+\rangle + \delta |-\rangle$
 - 1. First, diagonalize the measured "observable"

 $X = \left|+\right\rangle \left\langle+\right| - \left|-\right\rangle \left\langle-\right|$

2. Define projectors from eigenvectors

 $M_{+1} = \left|+\right\rangle \left\langle+\right| \,, \, M_{-1} = \left|-\right\rangle \left\langle-\right| \,$

3. Possible outcomes "+1", "-1"

$$\mathbb{P}[+1] = \langle \psi | M_{+1} | \psi \rangle = |\gamma|^2$$
$$\mathbb{P}[-1] = \langle \psi | M_{-1} | \psi \rangle = |\delta|^2$$





Projective measurement

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 $|0\rangle$

 $|1\rangle$

- Measure X on $|\psi\rangle = \alpha |0\rangle + \beta |1\rangle = \gamma |+\rangle + \delta |-\rangle$
 - 1. First, diagonalize the measured "observable"

 $X = \left|+\right\rangle \left\langle+\right| - \left|-\right\rangle \left\langle-\right|$

2. Define projectors from eigenvectors

 $M_{+1} = \left|+\right\rangle \left\langle+\right| \,, \, M_{-1} = \left|-\right\rangle \left\langle-\right| \,$

3. Possible outcomes "+1", "-1"

$$\mathbb{P}[+1] = \langle \psi | M_{+1} | \psi \rangle = |\gamma|^2$$
$$\mathbb{P}[-1] = \langle \psi | M_{-1} | \psi \rangle = |\delta|^2$$

4. Post-measurement state

$$|\psi_{\pm}\rangle = \frac{M_{\pm 1} |\psi\rangle}{\sqrt{\mathbb{P}[\pm 1]}} = |\pm\rangle$$







 $U_1 U_2 \left| \psi \right\rangle$

• Single-qubit gates $|\psi\rangle - U_2 - U_1$ P - P = -





• Single-qubit measurements

Universal Set



Poll Question 14

Let the initial state be $|0\rangle$. We apply the H gate and then measure in the Z basis. What is the probability of the measurement result -1 and what is the corresponding post-measurement state?

- A. $\frac{1}{2}$ and $|1\rangle$
- *B.* $\frac{1}{\sqrt{2}}$ and $|1\rangle$
- C. $\frac{1}{2}$ and $|0\rangle$
- *D.* $\frac{1}{\sqrt{2}}$ and $|0\rangle$
- E. I'm not sure





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Multiple Qubits







Moving beyond one qubit

Controlled-NOT gate: Flip target qubit if control qubit is 1

Control qubit $ a\rangle - a\rangle$	Input (<i>a</i> , <i>b</i>)	Output $(a, a \oplus b)$
Target qubit $ b angle = \bigoplus a \oplus b angle$	00	00
$\begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$	01	01
	10	11
$CNOT = \begin{bmatrix} 1 & 0 \\ 0 & X \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}$	11	10
$\begin{array}{c} CX \\ CX \\$		

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Universal gates on n qubits:



Kronecker (or Tensor) product



$$= \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{m1} & a_{m2} & \cdots & a_{mn} \end{bmatrix} \otimes \begin{bmatrix} b_{11} & b_{12} & \cdots & b_{1q} \\ b_{21} & b_{22} & \cdots & b_{2q} \\ \vdots & \vdots & \ddots & \vdots \\ b_{p1} & b_{p2} & \cdots & b_{pq} \end{bmatrix} \\ = \begin{bmatrix} a_{11} \times \begin{bmatrix} b_{11} & \cdots & b_{1q} \\ \vdots & \ddots & \vdots \\ b_{p1} & \cdots & b_{pq} \end{bmatrix} & a_{12} \times B & \cdots & a_{1n} \times B \\ a_{21} \times B & a_{22} \times B & \cdots & a_{2n} \times B \\ \vdots & \vdots & \ddots & \vdots \\ a_{m1} \times B & a_{m2} \times B & \cdots & a_{mn} \times B \end{bmatrix}_{x}$$

 $\exists mp \times nq$

Useful Property: $(A \otimes B)(C \otimes D) = (AC) \otimes (BD)$



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 $A_{m \times n} \otimes B_{p \times q}$





Moving beyond one qubit

Controlled-NOT gate: Flip target qubit if control qubit is 1

Control qubit	$ a\rangle$ —		\rangle		Input (<i>a</i> , <i>b</i>)	Output $(a, a \oplus b)$
Target qubit	$ b\rangle - \epsilon$	\mathbf{F} a	$\oplus b angle$		00	00
				0	01	01
$ 10\rangle = 1\rangle \otimes 0\rangle = \begin{bmatrix} 0\\1 \end{bmatrix} \otimes \begin{bmatrix} 1\\0 \end{bmatrix} = \begin{bmatrix} 0\\1 \end{bmatrix}$	10	11				
	11	10				
$\operatorname{CNOT} 10\rangle =$	$\begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1 \end{bmatrix}$	$ \begin{array}{ccc} 0 & 0 \\ 0 & 0 \\ 0 & 1 \\ 1 & 0 \end{array} $	$\begin{bmatrix} 0\\0\\1\\0\end{bmatrix}$	$\begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 1 \end{bmatrix}$	$= \begin{bmatrix} 0 \\ 1 \end{bmatrix} \otimes$	$\begin{bmatrix} 0\\1\end{bmatrix} = 11\rangle$

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Superposition + Linearity → Entanglement

$$|0\rangle - H + \rangle = \frac{|0\rangle + |1\rangle}{\sqrt{2}}$$
Superposition
$$|0\rangle - H + \rangle \otimes |0\rangle = \frac{|00\rangle + |10\rangle}{\sqrt{2}}$$
Add a qubit
$$|0\rangle - H + \rangle \otimes |0\rangle = \frac{|00\rangle + |10\rangle}{\sqrt{2}}$$
Linearity
$$|0\rangle - H + \int CNOT \left(\frac{|00\rangle + |10\rangle}{\sqrt{2}}\right) = \frac{|00\rangle + |11\rangle}{\sqrt{2}}$$
Linearity
Entanglement!

 $rac{|00
angle+|11
angle}{\sqrt{2}}$ CANNOT be expressed as a tensor product $|\psi
angle\otimes|\phi
angle$







• Computational basis states:

 $|\mathbf{v}\rangle = |v_1v_2\cdots v_n\rangle = |v_1\rangle \otimes |v_2\rangle \otimes \cdots \otimes |v_n\rangle \in \mathbb{C}^{2^n}; \ v_i \in \{0,1\}$

• State vector for an *n*-qubit state:

$$|\psi\rangle = \sum_{\mathbf{v}\in\{0,1\}^n} \alpha_{\mathbf{v}} \,|\mathbf{v}\rangle \in \mathbb{C}^{2^n} \;;\; \alpha_{\mathbf{v}}\in\mathbb{C}, \; \||\psi\rangle\|_2^2 = \sum_{\mathbf{v}\in\{0,1\}^n} |\alpha_{\mathbf{v}}|^2 = 1$$

- Unitary operations on the state: $|\psi\rangle \mapsto U |\psi\rangle \in \mathbb{C}^{2^n}$; $U \in \mathbb{U}^{2^n \times 2^n}, U^{-1} = U^{\dagger}, ||U |\psi\rangle||_2 = 1$
- Projective measurement of an "observable" *O*:

$$O = O^{\dagger} = \sum_{i} m_{i} M_{i} , \ \mathbb{P}[m_{i}] = \langle \psi | M_{i} | \psi \rangle , \ |\psi_{i}\rangle = \frac{M_{i} |\psi\rangle}{\sqrt{\mathbb{P}[m_{i}]}}$$

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Poll Question 15

What is the result of $(X \otimes Z)(|0\rangle \otimes |1\rangle)$?

- A. $|1\rangle\otimes|1\rangle$
- *B.* $-|1\rangle\otimes|0\rangle$
- $C. -|0
 angle \otimes |1
 angle$
- D. $-|1\rangle\otimes|1\rangle$
- E. I'm not sure





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Entanglement and Stabilizers





Entanglement: Bell Pairs





What happens if we measure the first qubit in the Z basis?

$$M_{+1} = |0\rangle \langle 0|_{\mathcal{A}} \otimes I_{\mathcal{B}}, \ M_{-1} = |1\rangle \langle 1|_{\mathcal{A}} \otimes I_{\mathcal{B}}$$

First qubit collapses to 0/1 & so does the second qubit too!

Bell Basis:
$$|\Phi^{\pm}\rangle_{AB} = \frac{|00\rangle_{AB} \pm |11\rangle_{AB}}{\sqrt{2}}$$
, $|\Psi^{\pm}\rangle = \frac{|01\rangle_{AB} \pm |10\rangle_{AB}}{\sqrt{2}}$



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Bell Basis:
$$|\Phi^{\pm}\rangle_{AB} = \frac{|00\rangle_{AB} \pm |11\rangle_{AB}}{\sqrt{2}}$$
, $|\Psi^{\pm}\rangle_{AB} = \frac{|01\rangle_{AB} \pm |10\rangle_{AB}}{\sqrt{2}}$

These are ± 1 -eigenvalued eigenvectors of ZZ, XX, -YY, II:

$$Z_{A}Z_{B} |\Phi^{\pm}\rangle_{AB} = (Z \otimes Z) \left(\frac{|0\rangle \otimes |0\rangle \pm |1\rangle \otimes |1\rangle}{\sqrt{2}} \right)$$
$$= \frac{Z |0\rangle \otimes Z |0\rangle \pm Z |1\rangle \otimes Z |1\rangle}{\sqrt{2}}$$
$$= \frac{|0\rangle \otimes |0\rangle \pm (-|1\rangle) \otimes (-|1\rangle)}{\sqrt{2}}$$
$$= |\Phi^{\pm}\rangle_{AB}$$









Bell Basis:
$$|\Phi^{\pm}\rangle_{AB} = \frac{|00\rangle_{AB} \pm |11\rangle_{AB}}{\sqrt{2}}$$
, $|\Psi^{\pm}\rangle_{AB} = \frac{|01\rangle_{AB} \pm |10\rangle_{AB}}{\sqrt{2}}$

These are ± 1 -eigenvalued eigenvectors of ZZ, XX, -YY, II:

$$Z_{A}Z_{B} |\Psi^{\pm}\rangle_{AB} = (Z \otimes Z) \left(\frac{|0\rangle \otimes |1\rangle \pm |1\rangle \otimes |0\rangle}{\sqrt{2}} \right)$$
$$= \frac{Z |0\rangle \otimes Z |1\rangle \pm Z |1\rangle \otimes Z |0\rangle}{\sqrt{2}}$$
$$= \frac{|0\rangle \otimes (-|1\rangle) \pm (-|1\rangle) \otimes |0\rangle}{\sqrt{2}}$$
$$= -|\Psi^{\pm}\rangle_{AB}$$









Stabilizer States

Elements of the stabilizer must mutually commute to have a common eigenbasis, i.e., the same set of eigenvectors diagonalize all stabilizer elements. Key fact: XZ = -ZX

An n-qubit stabilizer state has n Pauli stabilizer generators











GHZ Basis: $|GHZ\rangle_{ABC} = \frac{|000\rangle_{ABC} + |111\rangle_{ABC}}{\sqrt{2}}$ and its variants (GHZ: Greenberger-Horne-Zeilinger)

Stabilizers: $\langle \pm Z_A Z_B I_C, \pm I_A Z_B Z_C, \pm X_A X_B X_C \rangle$

An n-qubit stabilizer state has n Pauli stabilizer generators



Poll Question 16

What are the stabilizer generators of $\frac{|001\rangle + |110\rangle}{\sqrt{2}}$? The state must have eigenvalue

- + 1 for these operators.
- A. $\langle ZZI, IZZ, XXX \rangle$
- B. $\langle ZZI, IZZ, -XXX \rangle$
- C. $\langle ZZI, -IZZ, XXX \rangle$
- $D. \ \langle -ZZI, IZZ, XXX \rangle$
- *E.* $\langle ZZI, -IZZ, -XXX \rangle$
- $F. \quad \langle -ZZI, -IZZ, XXX \rangle$
- G. I'm not sure





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The Stabilizer Formalism









Here we can track the state quite easily, with length 4 vectors



This is more complicated to track! With n qubits we have length 2^{n} !









Unitary operations U that map Paulis to Paulis under conjugation $C_n = \{U \in \mathbb{U}^{2^n} : UEU^{\dagger} = E' \in \mathcal{P}_n \text{ for all } E \in \mathcal{P}_n\}$

 $\mathcal{P}_n = \{\pm i E_1 \otimes E_2 \otimes \cdots \otimes E_n ; E_j \in \{I, X, Y, Z\}, j = 1, 2, \dots, n\}$ (Clifford group & Pauli group)







Clifford gates: Pauli tracking













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Pauli measurements





Compare measured operator Z_A with each stabilizer

- 1. Z_A anticommutes with $X_A X_B$: replace with $\pm Z_A$
- 2. Z_A commutes with $Z_A Z_B$: retain the stabilizer

Output stabilizer: $\langle \pm Z_A , Z_A Z_B \rangle \equiv \langle \pm Z_A , \pm Z_B \rangle \equiv |00\rangle/|11\rangle$





Pauli measurements





Compare measured operator Y_A with each stabilizer

- 1. Y_A anticommutes with $X_A X_B$: replace with $\pm Y_A$
- 2. Y_A anticommutes with $Z_A Z_B$: multiply $Z_A Z_B$ with $X_A X_B$

Output stabilizer: $\langle \pm Y_A, -Y_A Y_B \rangle \equiv \langle \pm Y_A, \mp Y_B \rangle$



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Clifford gates and Pauli measurements on input stabilizer states can be efficiently simulated classically, by simply tracking the stabilizers of the input state through the circuit!

Stabilizer Circuits: Cliffords + Pauli measurements


Poll Question 17

What are the stabilizers for the output of the following circuit if the measurement result is +1?



- $A. \langle -XX, XI \rangle = \langle -IX, XI \rangle$ $B. \langle -XX, -XI \rangle = \langle IX, -XI \rangle$ $C. \langle -XX, ZZ \rangle = \langle -XX, YY \rangle$ $D. \langle XX, -XI \rangle = \langle -IX, -XI \rangle$
- E. I'm not sure





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Protecting information with entanglement









Pauli errors (and beyond)

• Recall the *n*-qubit Pauli group:

 $\mathcal{P}_n = \{ \pm i E_1 \otimes E_2 \otimes \cdots \otimes E_n ; E_j \in \{I, X, Y, Z\}, j = 1, 2, \dots, n \}$

- Each element can also be thought of as an error operator, since Pauli matrices form an orthogonal basis for all matrices under the trace inner product: $\langle A, B \rangle_{Tr} := Tr(A^{\dagger}B)$
- Key Result: if Pauli errors on *t* qubits can be corrected, then any linear combination of them can also be corrected
- Goal: design quantum codes that correct Pauli errors







Bell Basis:
$$|\Phi^{\pm}\rangle_{AB} = \frac{|00\rangle_{AB} \pm |11\rangle_{AB}}{\sqrt{2}}$$
, $|\Psi^{\pm}\rangle_{AB} = \frac{|01\rangle_{AB} \pm |10\rangle_{AB}}{\sqrt{2}}$
(EPR: Einstein-Podolsky-Rosen)
Stabilizers: $\langle Z_A Z_B , \pm X_A X_B \rangle$ $\langle -Z_A Z_B , \pm X_A X_B \rangle$

GHZ Basis: $|GHZ\rangle_{ABC} = \frac{|000\rangle_{ABC} + |111\rangle_{ABC}}{\sqrt{2}}$ and its variants (GHZ: Greenberger-Horne-Zeilinger)

Stabilizers: $\langle \pm Z_A Z_B I_C , \pm I_A Z_B Z_C , \pm X_A X_B X_C \rangle$

An n-qubit stabilizer state has n Pauli stabilizer generators







The three-qubit code





From GHZ stabilizers $\langle ZZI, IZZ, XXX \rangle$ drop XXX to create a logical qubit! Stabilizers: $\langle ZZI, IZZ \rangle$ (they commute), $\overline{|\psi\rangle}$ is a +1-eigenvector for all α, β





Syndrome measurement



Suppose that after encoding the logical qubit the error X_1 acts on the state



Measure the stabilizer generators $S_1 = ZZI$ and $S_2 = IZZ$:









The error *X* propagates through the CNOT and flips the measurement

Hence, the measurement results in -1 whenever there are an odd number of X's on the ancilla (through the CNOT gates), i.e., when the error anticommutes with the stabilizer S_i

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Poll Question 18

Given the stabilizers $\langle S_1 = ZZI, S_2 = IZZ \rangle$ of the code, what is the syndrome for the error *IXI*?

- *A.* (+1, +1)
- *B.* (+1,−1)
- *C.* (-1, +1)
- *D.* (−1, −1)
- E. I'm not sure





Binary representation



Map an *n*-qubit Hermitian Pauli matrix to a pair of binary vectors:

Example for
$$n = 3$$
: $X \otimes Z \otimes Y \longrightarrow E(a, b)$ $a = \begin{bmatrix} 1 & 0 & 1 \end{bmatrix}$ (X component) $b = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix}$ (Z component)

How to check if $E(a, b) = X \otimes Z \otimes Y$ and $E(c, d) = Z \otimes Z \otimes X$ commute?

Compare operators on each qubit: $X \otimes Z \otimes Y \mapsto ([1 \ 0 \ 1], [0 \ 1 \ 0])$ $Z \otimes Z \otimes X \mapsto ([0 \ 0 \ 1], [1 \ 1 \ 0])$

Symplectic inner product: $\langle [a, b], [c, d] \rangle_{sym} \coloneqq ad^T + bc^T$ (modulo 2) = $\begin{cases} 0 & \text{iff they commute,} \\ 1 & \text{iff they anticommute} \end{cases}$



Stabilizers
$$(n = 3)$$
:
 $Z \otimes Z \otimes I$
 $I \otimes Z \otimes Z$

 (X component)
 $a_1 = [0 \quad 0 \quad 0]$
 $a_2 = [0 \quad 0 \quad 0]$

 (Z component)
 $b_1 = [1 \quad 1 \quad 0]$
 $b_2 = [0 \quad 1 \quad 1]$

Let the error operator be $X \otimes I \otimes I \equiv XII = E(c, d) = E([1 \ 0 \ 0], [0 \ 0 \ 0])$

Symplectic inner product: $\langle [a, b], [c, d] \rangle_{sym} \coloneqq ad^T + bc^T$ (modulo 2)

Syndrome =
$$\begin{bmatrix} \langle [\boldsymbol{a}_1, \boldsymbol{b}_1], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \\ \langle [\boldsymbol{a}_2, \boldsymbol{b}_2], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

 $a_1d^T + b_1c^T = [0\ 0\ 0]\ [0\ 0\ 0]^T + [1\ 1\ 0]\ [1\ 0\ 0]^T = 0 + 1 = 1 \pmod{2}$



Stabilizers (n = 3):

$$Z \otimes Z \otimes I$$
 $I \otimes Z \otimes Z$

 (X component) $a_1 = [0 \quad 0 \quad 0]$
 $a_2 = [0 \quad 0 \quad 0]$

 (Z component) $b_1 = [1 \quad 1 \quad 0]$
 $b_2 = [0 \quad 1 \quad 1]$

$$H = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a \mid H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Symplectic inner product: $\langle [a, b], [c, d] \rangle_{sym} \coloneqq ad^T + bc^T$ (modulo 2)

Syndrome =
$$\begin{bmatrix} \langle [\boldsymbol{a}_1, \boldsymbol{b}_1], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \\ \langle [\boldsymbol{a}_2, \boldsymbol{b}_2], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \end{bmatrix} = H_a d^T + H_b c^T$$



Minimum Distance and Logical Operators

$$H = \begin{bmatrix} 0 & 0 & 0 & | & 1 & 1 & 0 \\ 0 & 0 & 0 & | & 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & | & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Syndrome =
$$\begin{bmatrix} \langle [\boldsymbol{a}_1, \boldsymbol{b}_1], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \\ \langle [\boldsymbol{a}_2, \boldsymbol{b}_2], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \end{bmatrix} = H_a d^T + H_b c^T$$

What are the "codewords" of this quantum code? Generated by $\overline{X} = [1\ 1\ 1\ ,0\ 0\ 0]$ and $\overline{Z} = [0\ 0\ 0\ ,1\ 0\ 0]$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators



Poll Question 19

Given the parity-check matrix $H = \begin{bmatrix} 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 \end{bmatrix}$ of the code, what is the binary syndrome for the error *XXX*?

- *A.* [0,0]^{*T*}
- *B.* $[0,1]^T$
- *C.* $[1,0]^T$
- *D.* $[1,1]^T$
- E. I'm not sure







$$H = \begin{bmatrix} 0 & 0 & 0 & | & 1 & 1 & 0 \\ 0 & 0 & 0 & | & 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & | & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators







$$H = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators

Generated by
$$\overline{X} = [1 \ 1 \ 1 \ , 0 \ 0 \ 0]$$
 and $\overline{Z} = [0 \ 0 \ 0 \ , 1 \ 0 \ 0]$

$$\begin{aligned} \left|\psi\right\rangle_{L} &= \alpha \left|0\right\rangle + \beta \left|1\right\rangle \underbrace{\frac{Z}{\left|0\right\rangle}}_{\left|0\right\rangle} \underbrace{\frac{Z}{\left|0\right\rangle}}_{\left|0\right\rangle} \left|\overline{\psi}\right\rangle &= \alpha \left|000\right\rangle + \beta \left|111\right\rangle \end{aligned}$$







$$H = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators

Generated by
$$\overline{X} = [1 \ 1 \ 1 \ , 0 \ 0 \ 0]$$
 and $\overline{Z} = [0 \ 0 \ 0 \ , 1 \ 0 \ 0]$

$$\begin{aligned} |\psi\rangle_L &= \alpha |0\rangle + \beta |1\rangle & \overbrace{I} \\ & |0\rangle & \overbrace{I} \\ & |0\rangle & \overbrace{I} \end{aligned} \\ \begin{bmatrix} 3,1,1 \end{bmatrix} Code \\ \hline |\psi\rangle &= \alpha |000\rangle + \beta |111\rangle \end{aligned}$$







$$H = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators

$$\begin{array}{c} X \left|\psi\right\rangle_{L} = \alpha \left|0\right\rangle + \beta \left|1\right\rangle & \longrightarrow \\ & \left|0\right\rangle & \bigoplus \end{array} \end{array} \right\} \overline{\left|\psi\right\rangle} = \alpha \left|000\right\rangle + \beta \left|111\right\rangle \\ & \left|0\right\rangle & \bigoplus \end{array} \right\}$$







$$H = \begin{bmatrix} 0 & 0 & 0 & | & 1 & 1 & 0 \\ 0 & 0 & 0 & | & 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & | & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators

$$\begin{aligned} |\psi\rangle_{L} &= \alpha |0\rangle + \beta |1\rangle \xrightarrow{X} \\ |0\rangle \xrightarrow{X} \\ |0\rangle \xrightarrow{X} \\ |0\rangle \xrightarrow{X} \\ |\psi\rangle = \alpha |000\rangle + \beta |111\rangle \end{aligned}$$







$$H = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} H_a & H_b \end{bmatrix}; \quad H_a H_b^T + H_b H_a^T = \mathbf{0}$$

Minimum Distance: Minimum weight of any codeword Codewords are referred to as logical operators

$$\begin{aligned} |\psi\rangle_L &= \alpha |0\rangle + \beta |1\rangle \xrightarrow{X} \\ |0\rangle \xrightarrow{X} \\ |0\rangle \xrightarrow{X} \end{aligned} \Big\{ \overline{|\psi\rangle} &= \alpha |000\rangle + \beta |111\rangle \\ \hline |\psi\rangle &= \alpha |000\rangle + \beta |111\rangle \end{aligned}$$







[7,4,3] Hamming Code:
$$H = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 1 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \end{bmatrix}$$

[[7,1,3]] Steane Code: $H_S = \begin{bmatrix} H_a & H_b \end{bmatrix} = \begin{bmatrix} H & 0 \\ \hline 0 & H \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$

$$H_a H_b^T + H_b H_a^T = \begin{bmatrix} 0 & H H^T \\ \hline 0 & 0 \end{bmatrix} = 0$$



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 2^k



ML decoding:	$\hat{x} = \operatorname{argmin} d(x, y = 110101$	$\Big)$
	$x{\in}\mathcal{C}$	

	000 000	001 011	010 110	100 101	011 101	101 110	110 011	111 000	00
	000 001	001 010	010 111	100 100	011 100	101 111	110 010	111 001	00
	000 010	001 001	010 100	100 111	011 111	101 100	110 001	111 010	01
$\overset{\circ}{\prec}$	000 100	001 111	010 010	100 001	011 001	101 010	110 111	111 100	10
	001 000	000 011	011 110	101 101	010 101	100 110	111 011	110 000	01
	010 000	011 011	000 110	<u>110 101</u>	001 101	111 110	100 011	101 000	11
	100 000	101 011	110 110	000 101	111 101	001 110	010 011	011 000	10

Complexity scales exponentially!!





[[7,1,3]] Steane Code:
$$H_S = \begin{bmatrix} H_a & H_b \end{bmatrix} = \begin{bmatrix} H & 0 \\ \hline 0 & H \end{bmatrix} \begin{bmatrix} X \\ Z \end{bmatrix}$$

Syndrome =
$$\begin{bmatrix} \langle [\boldsymbol{a_1}, \boldsymbol{b_1}], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \\ \vdots \\ \langle [\boldsymbol{a_{14}}, \boldsymbol{b_{14}}], [\boldsymbol{c}, \boldsymbol{d}] \rangle_{\text{sym}} \end{bmatrix} = H_a d^T + H_b c^T$$

Syndrome decoding: Given the measured syndrome, determine the most likely error [c, d] that matches the measured syndrome

Complexity scales exponentially!!



Poll Question 20

Given the Steane code parity-check matrix $H_S = [H_a \mid H_b] = \begin{bmatrix} H & 0 \\ 0 & H \end{bmatrix}$ with $H = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 1 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \end{bmatrix},$

what is the most likely error pattern corresponding to the syndrome $[0,1,0,0,0,0]^T$? Assume that the channel is memoryless, so it applies independent Pauli errors.

- $A. \quad [0\ 0\ 0\ 0\ 0\ 0\ 0\ |\ 1\ 0\ 0\ 1\ 0\ 0\ 0]$

- $E. \quad \begin{bmatrix} 0 \ 1 \ 0 \ 0 \ 0 \ 0 \ 0 \end{bmatrix} | 0 \ 0 \ 0 \ 0 \ 0 \ 0 \end{bmatrix}$
- F. I'm not sure



CSS (Calderbank-Shor-Steane) Codes

• Consider two classical codes C_X and C_Z whose paritycheck matrices H_X and H_Z satisfy $H_X H_Z^T = 0$

- Define the CSS (stabilizer) code by $H_{\text{CSS}} = \begin{bmatrix} H_X & 0 \\ 0 & H_Z \end{bmatrix}$
- Logical operators [*c*, *d*] defined by $H_X d^T + H_Z c^T = 0$
- Error $[e_X, e_Z] \Rightarrow$ syndrome is $s = H_X e_Z^T + H_Z e_X^T \pmod{2}$
- $[[n, k, d]] = [[n, k_X + k_Z n, w_{\min}([C_X \setminus C_Z^{\perp}] \cup [C_Z \setminus C_X^{\perp}])]]$

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Surface code



 $[[O(L^2), 1, L]]$



- \Box plaquette checks (H_Z)
- - Logical Z
- Logical X



Wang et al. http://arxiv.org/abs/0905.0531

Poll Question 21

Consider the following statements and answer if they are true or false:

- 1. Errors that produce a zero syndrome must be stabilizers or logical operators
- 2. The surface code stabilizer generators each involve either 3 or 4 qubits
- A. 1 is True, 2 is False
- B. 1 is False, 2 is True
- C. 1 is True, 2 is True
- D. 1 is False, 2 is False
- E. I'm not sure







Tucson Arizona

01/05/2023







• Consider two classical LDPC codes C_X and C_Z whose parity-check matrices H_X and H_Z satisfy $H_X H_Z^T = 0$

• Define the CSS QLDPC code by $H_{\text{QLDPC}} = \begin{bmatrix} H_X & 0 \\ 0 & H_Z \end{bmatrix}$

- Several QLDPC code families exist:
 - Hypergraph Product codes, e.g., the surface code
 - Bicycle and Generalized Bicycle codes
 - Homological Product codes
 - Lifted Product codes
 - Quantum Tanner codes



Syndrome-based iterative decoding



Belief propagation (BP)



Variable node (VN) update:

$$\mu_{x \to f}(x) = \prod_{h \in n(x) \setminus \{f\}} \mu_{h \to x}(x)$$

Check node (CN) update:

$$\mu_{f \to x}(x) = \sum_{\sim \{x\}} \left(f(X) \prod_{h \in n(f) \setminus \{x\}} \mu_{y \to f}(y) \right)$$

Variable node (VN) decision:

$$g_i(x_i) = \prod_{h \in n(x_i)} \mu_{h \to x_i}(x_i)$$



Poll Question 22

Consider a CSS QLDPC code constructed from classical codes C_X and C_Z . Then which of the following is false?

- A. H_X and H_Z are orthogonal
- *B.* H_X and H_Z are sparse, i.e., have very few 1s
- C. The code is a stabilizer code
- D. Any stabilizer code is a CSS code
- E. Universal computation requires fault-tolerant realizations of *H*, *T*, *CNOT* on the logical qubits
- F. I'm not sure



Error Correction: Classical vs Quantum

Classical: Decode based on received vector
 Quantum: Decode based only on measured syndrome

 Classical: Any sparse parity-check matrix gives LDPC Quantum: Need two sparse matrices that are orthogonal

- Classical: Only the zero vector causes trivial syndrome Quantum: All stabilizers have zero syndrome (degeneracy)
- Classical: Hardware noise quite low, mainly channel noise Quantum: Everything noisy – decoding + logical gates







- How to fully leverage degeneracy in QLDPC decoders?
- Local iterative algorithms that correct many errors?
- Can we physically realize good QLDPC codes in hardware despite their many long-range connections?
- Universal fault-tolerance on good QLDPC codes?
- ... and many more!



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Course Evaluation Survey

We value your feedback on all aspects of this short course. Please go to the link provided in the Zoom Chat or in the email you will soon receive to give your opinions of what worked and what could be improved.

CQN Winter School on Quantum Networks

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